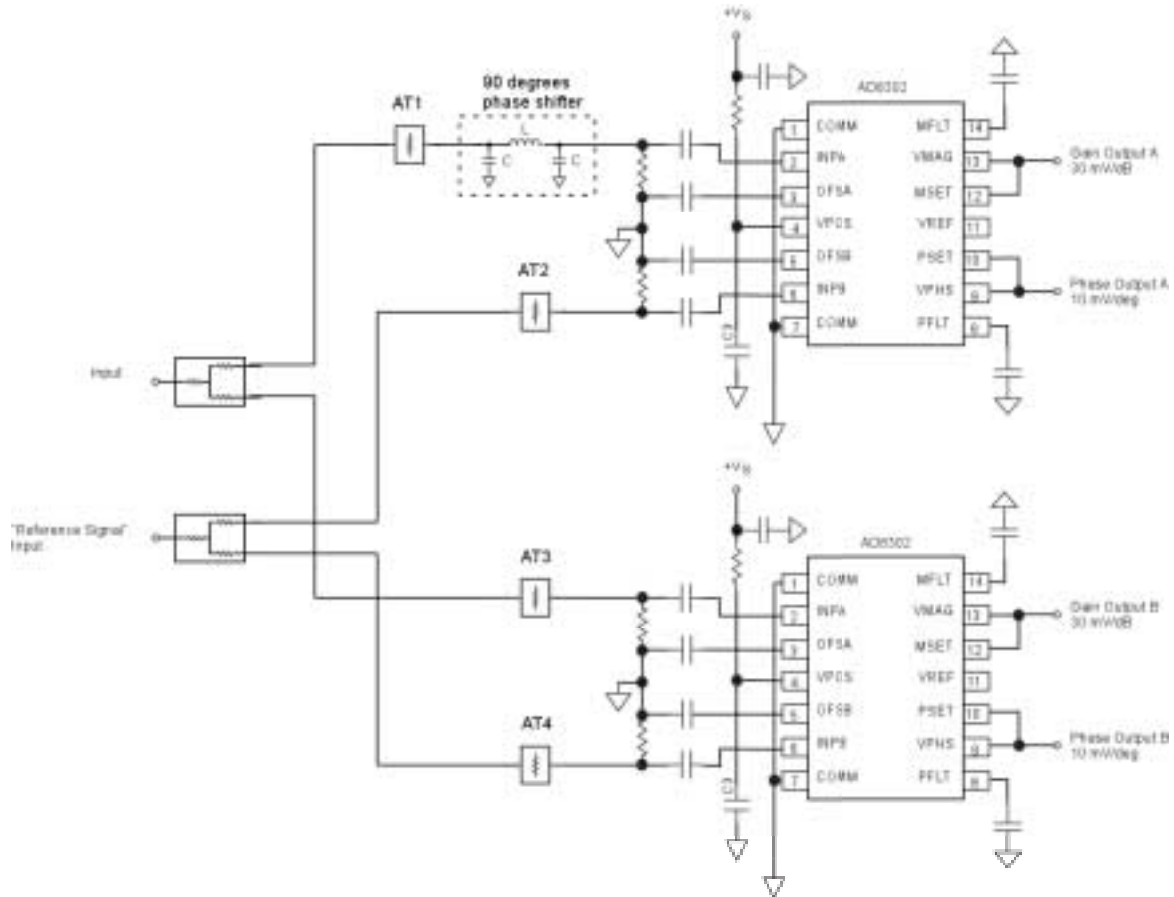
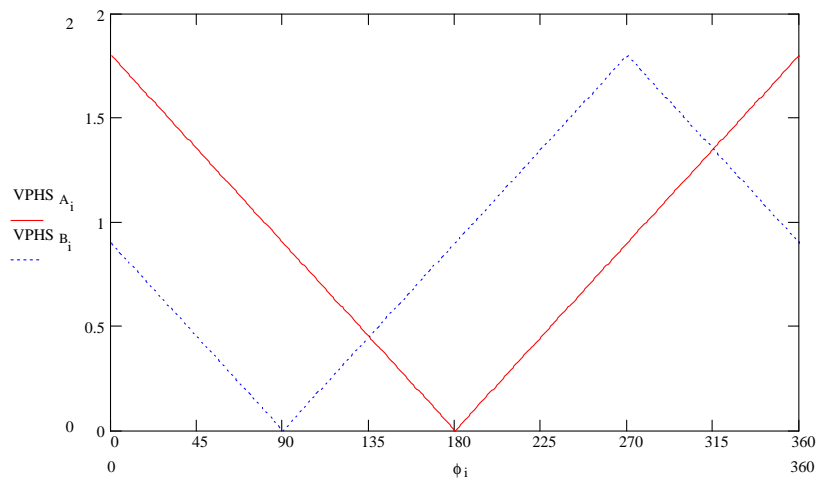


## a Full 360° Phase Detection with AD8302 180° Phase Detectors

1. use two AD8302's
  - a. apply identical "reference signal" to each AD8302 INPB
  - b. apply "unknown signal" to INPA of one of the AD8302's
  - c. apply 90°-shifted "unknown signal" to INPA of other AD8302



2. Resulting VPHS outputs are shown below



## a Full 360° Phase Detection with AD8302 180° Phase Detectors

3. Determine phase angle,  $\Theta$ , through following decision tree:

- a. If  $V_{PHS_A} \geq (0.75 * V_{REF})$  and  $(V_{PHS_B} \leq V_{REF}/2)$  then  $\Theta = ((V_{PHS_B} - 0.9)/-10^{-2})$
- b. If  $V_{PHS_B} < (0.25 * V_{REF})$  then  $\Theta = ((V_{PHS_A} - 0.9)/-10^{-2}) + 90$
- c. If  $V_{PHS_A} \leq (0.25 * V_{REF})$  then  $\Theta = ((V_{PHS_B} - 0.9)/10^{-2}) + 180$
- d. If  $V_{PHS_B} > (0.75 * V_{REF})$  then  $\Theta = ((V_{PHS_A} - 0.9)/10^{-2}) + 270$
- e. If  $V_{PHS_A} \geq (0.75 * V_{REF})$  and  $(V_{PHS_B} \geq V_{REF}/2)$  then  $\Theta = ((V_{PHS_B} - 0.9)/-10^{-2}) + 360$

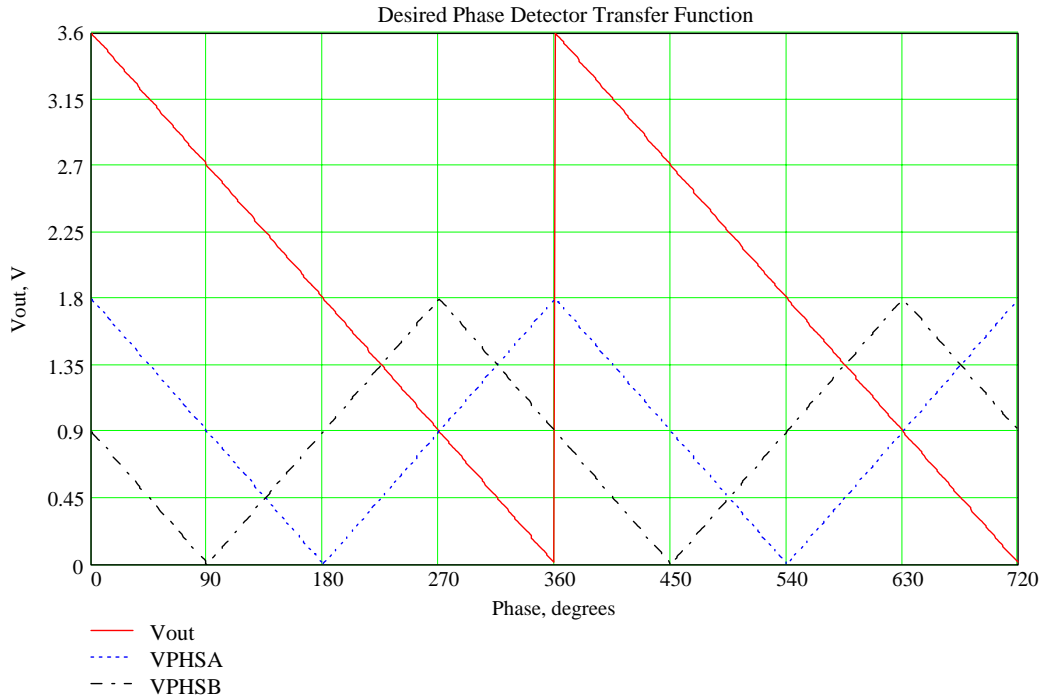
(Note that there are “greater than” ( $>$ ), “greater than or equal to” ( $\geq$ ), “less than” ( $<$ ) and “less than or equal to” ( $\leq$ ) inequalities in the decision tree above).

This algorithm could be implemented by digitizing  $V_{PHS_A}$  and  $V_{PHS_B}$  and using an IF-THEN-ELSE or similar structure in software, then either use this digital value directly or use this digital value to produce an analog output with a DAC.

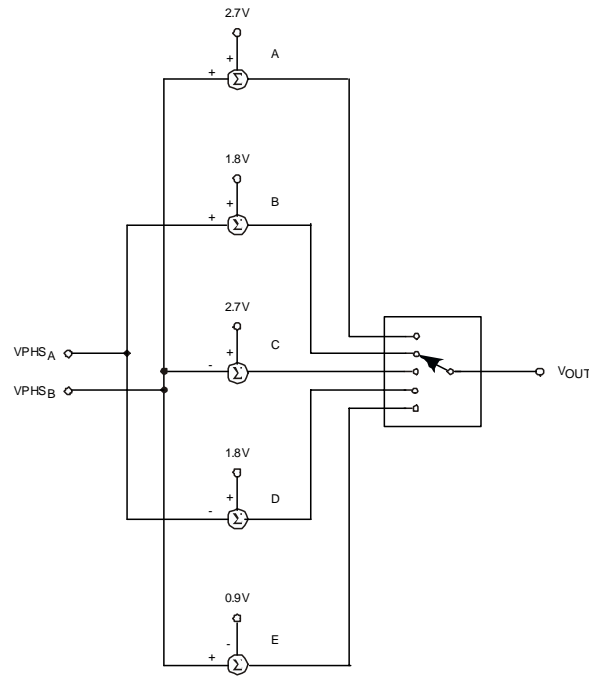
This algorithm uses only the “sweet spot” regions of each  $V_{PHS}$  output, i.e., those portions of the curves that are centered around 900 mV.

4. Analog output voltage could be constructed from  $V_{PHS_A}$  and  $V_{PHS_B}$  according to following decision tree:
  - a. If  $\Theta \leq 45^\circ$  then  $V_{OUT} = V_{PHS_B} + 2.7 \text{ V}$
  - b. If  $(\Theta > 45^\circ)$  and  $(\Theta \leq 135^\circ)$  then  $V_{OUT} = V_{PHS_A} + 1.8 \text{ V}$
  - c. If  $(\Theta > 135^\circ)$  and  $(\Theta \leq 225^\circ)$  then  $V_{OUT} = 2.7 \text{ V} - V_{PHS_B}$
  - d. If  $(\Theta > 225^\circ)$  and  $(\Theta \leq 315^\circ)$  then  $V_{OUT} = 1.8 \text{ V} - V_{PHS_A}$
  - e. If  $\Theta > 315^\circ$  then  $V_{OUT} = V_{PHS_B} - 0.9 \text{ V}$

# a Full 360° Phase Detection with AD8302 180° Phase Detectors



This waveform could be constructed using the VREF output to produce the constants ( $3/2 V_{REF}$ ,  $V_{REF}$ ,  $1/2 V_{REF}$ ) that are combined with the appropriate VPHS output using op amp summer and subtractor circuits as shown below



Voltage A is selected when the phase angle,  $\Theta$ , is  $> 0^\circ$  and  $\leq 45^\circ$ . Voltage B is selected when  $\Theta$  is  $> 45^\circ$  but  $\leq 135^\circ$ , Voltage C is selected for  $135^\circ < \Theta \leq 225^\circ$ , Voltage D for

## a Full 360° Phase Detection with AD8302 180° Phase Detectors

$225^\circ < \Theta \leq 315^\circ$  and Voltage E for  $315^\circ < \Theta \leq 360^\circ$ . The selection circuit, which could be a SP5T switch, 1:8 multiplexer, etc., would be controlled by comparators which would determine the proper output voltage selection by comparing  $VPHS_A$  and/or  $VPHS_B$  to  $0.75 * VREF$ ,  $0.25 * VREF$  and each other. This discrete-switching approach might be inferior to a soft-handoff approach that would gradually select transfer the input of the selection circuit.

The summing and subtracting circuits could be constructed from op amp circuits.